A High Speed Pipelined Dynamic Circuit Implementation Using Modified TSPC Logic Design Style With Improved Performance

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Abstract— The high-speed dynamic True Single Phase Clock (TSPC) logic design style offer fully pipelined logic circuits using only one clock signal, which makes clock distribution simple and compact. The conversion of simple logic gates to pipelined TSPC logic gates increases transistor count since standard cell implementation for a logic function uses both N-block as well as P-block to remove transparency between pipelined stages, despite the fact that logic functions are only implemented with N-block. In this paper we present a technique in which a TSPC logic cell are implemented both as cell_N and cell_P cells, where each cell block is performing a logic function along with only one type latching operation. Such an implementation allows a systematic approach for converting un-pipelined circuits to fully pipelined circuits. The alternate cell_N and cell_P behaves as dynamic register and removes transparency between pipelined stages. The appropriate numbers of dynamic registers are used to equalize stage delays for all paths and to remove transparency between pipelined stages. The modified TSPC implementation shows almost 40% to 50% reduction in transistor counts and almost 50% reduction in clock cycles as compared to worst-case standard TSPC implementation. The worst-case standard TSPC implementation assumes that no logic merging is possible with P-block, since input to any cell appears after different cycle delays. The modified TSPC logic circuit implementation preserves all the advantages of standard TSPC logic implementation and in addition offers the reduced circuit complexity due to reduced transistor count per logic cell. The proposed logic design style reduces layout area and average power consumption as compared to the standard TSPC pipelined circuit implementation.

Index Terms—TSPC, dynamic, pipelined, high speed, transistor count, circuit complexity

I. INTRODUCTION

Conventional static CMOS circuit implementation cannot fulfill the requirements of high throughput pipelined digital designs therefore dynamic logic circuits can be used to speed-up digital designs. The high-speed dynamic True Single Phase Clock (TSPC) logic design style offer fully pipelined logic circuits using only one clock signal, which makes clock distribution simple and compact [1], [2], [8]. The simple clock distribution of the TSPC circuits reduces the layout area for clock lines, no

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clock skew problem and increases the speed [1], [2], [8]. The TSPC Logic style shows a robust cell characteristic and therefore the standard cell implementation is possible. The TSPC-cell consists of one N-block and one P-block each driven by single clock signal (ϕ) [1], [2], [5], [7]. In standard TSPC logic style the implementation of simple gates like AND, OR, XOR increases the transistor count since each logic cell implementation use both N-block together with P-block to remove transparency between the pipelined stages, while logic function is implemented only with N-block [3], [5]; therefore it makes the P-block redundant, since it perform no logic function other than latching. Although logic merging is possible [1], [2] in standard TSPC cell by implementing logic functions with both the N-block and P-block, but such a logic merging is inefficient, since in practical cases inputs to a gate may appear after different cycle delays, which may not allow effective logic merging. The implementation complexity also increases due to reduced freedom for logic merging; since it requires the careful and tedious mapping of logic functions with both the blocks. The worst-case standard TSPC logic implementation assumes that each cell logic function is implemented with N-block only, while Pblock do not implement any logic but provide dynamic latching to avoid the transparency between pipelined stages (i.e. no logic merging possible). Furthermore, the pipelined circuit requires the stage registers (REG) to equalize the stage delays for all paths, which are implemented as one LATCH_N and LATCH_P in cascade. In standard TSPC logic cells if logic merging is not efficient, it increases pipelined stages that need the extra register operation, which increases the transistor count [3], [6]. In this paper we present a technique in which a TSPC logic cells are implemented as cell N and cell_P, where each cell is performing a logic function along with only one type latching operation (either Nlatch or P-latch). The transparency between the pipelined stages can be removed by alternate cell_N and cell_P assignment. Since each cell performs a logic operation along with latching, such an implementation will reduce the number of redundant transistors due to efficient merging of logic.

The circuit complexity gets reduced due to decrease in transistor count per logic cell. Since each cell performs the logic function, therefore the burden of stage register operations to equalize the stage delays for all paths gets



reduced by almost 50% as compared to worst-case standard TSPC logic implementation. The proposed technique reduces the latency and increases the throughput of digital pipelined circuit implementation since the number of clock cycle gets reduced by almost 50%. The proposed logic design style offers advantages like pipelining, low average power, low latency and high throughput. A 3-bit pipelined adder is designed using modified TSPC logic style and compared with standard **TSPC** implementation. logic circuit implementation using the proposed logic design style is described in Section II; circuit implementation is discussed in Section III. Section IV shows the comparison between standard TSPC and modified TSPC and section V concludes the paper.

II. LOGIC IMPLEMENTATION

The proposed modified TSPC implementation allows the flexibility of implementing logic function with both N-block as well as with P-block, hence converting the simpler logic cells to pipelined logic cells shows large reduction in the transistor count as compared to standard TSPC logic design style. The simple logic cells like AND, OR, XOR can be designed using both N-block logic cell as well as P-block logic cell, thus the logic cells produced for AND function will be AND_N & AND_P; OR, XOR in similar way produces OR_N & OR_P, XOR_N & XOR_P respectively. A simple AND2 gate can be implemented in two different ways (AND2_N and AND2 P) as shown in Figure 1. The operations of AND_N and AND2_P are shown in table I. REG can be split into LATCH_N and LATCH_P. The latching can be performed by these two latches are shown in Figure 2. These latches are used to equalize cycle delays, since the alternate LATCH_N and LATCH_P form a dynamic register REG [4]. Table II explains operation of both the latches. For a pipelined circuit implementation the cell-N and cell-P must be kept one after other in cascaded manner with appropriate latching to equalize stage delays for all paths as explained in section III. The operation of cell N/cell P and LATCH N/LATCH P are summarized in table III.

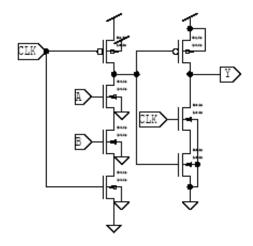
TABLE I. OPERATION OF AND_N AND OPERATION OF AND_P

OPERATION OF AND_N

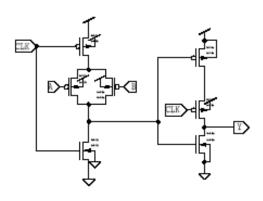
A	В	ф	Y	Comments
0	0	0	0	Dynamic evaluation
0	1	0	0	Dynamic evaluation
1	0	0	0	Dynamic evaluation
1	1	0	1	Dynamic evaluation
X	X	1	Latch	Dynamic latch

OPERATION OF AND_P

A	В	ф	Y	Comments
0	0	1	0	Dynamic evaluation
0	1	1	0	Dynamic evaluation



(a) AND_N



(b) AND_P

TABLE II. OPERATIONS OF LATCH_N AND LATCH_P

OPERATION OF LATCH_N

A	ф	Y	Comment
0	1	0	N1 is weak ON
1	1	1	P1 strong ON
0	0	Latch	Dynamic latch
1	0	Latch	Dynamic latch

OPERATION OF LATCH_P

A	ф	Y	Comment
0	0	0	N2 is strong ON
1	0	1	P2 weak ON
0	1	Latch	Dynamic latch
1	1	Latch	Dynamic latch



TABLE III. SUMMARY OF OPERATION

Cell Type	ф=1	ф=0
cell_N/	Dynamic	Dynamic
LATCH_N	evaluation	latch
cell_P/	Dynamic	Dynamic
LATCH_P	latch	evaluation

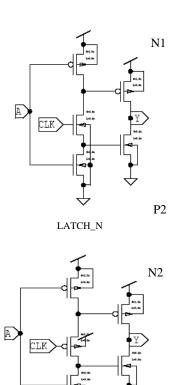


Figure 2. LATCH_N and LATCH_P Cells (CLK=\$)

LATCH_P

been designed using XOR2, AND2 and OR3 gates, where the first stage is implemented in cell_N logic while second stage is implemented as cell_P logic. A half cycle latch LATCH_N is used in parallel with cell_N logic to equalize stage delay, thus the circuit will take one clock cycle to produce correct output. In the standard TSPC implementation XOR2_N and XOR2_P are replaced by XOR2, LATCH_N is replaced by REG, AND_N is replaced by AND2, and OR3_P is replaced by OR3 therefore it takes 2 clock cycle to produce correct output. Comparison of transistor counts in two different implementations is shown in table IV. Following the hierarchical design approach, a 3-bit ripple pipelined adder can be designed using above 1-bit adder and keeping latches at appropriate positions as shown in Figure 4. The circuit takes three (cell_N+cell_P) delay; since one (cell_N+cell_P) forms dynamic register (REG) and takes one clock cycle to produce output therefore output would be available after 3-clock cycle delay for any applied input. The standard TSPC implementation uses REG instead of LATCH_N or LATCH_P and therefore requires almost double transistor and 6-clock cycle delay for any applied input. The comparison of transistor counts in two different implementations is shown in table IV. Both the circuit implementations were verified using transistor level simulator (T-spice) from Tanner Research Incorporation utilizing parameters of 0.5 μm technology. The simulation result shows correct pipelined circuit behavior. The supply voltage $V_{\rm DD}$ was kept at 3.3V.

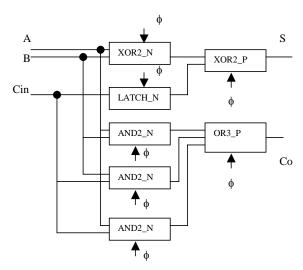


Figure 3. 1-bit pipelined adder circuit

IV. COMPARISON

The comparison of transistor counts for 3-bit pipelined adder implementation results utilizing standard TSPC logic design style and modified TSPC logic design style is shown in Table IV. The comparison result shows almost 46 to 47 % less transistor count for modified TSPC logic design style. The other advantage in modified TSPC logic implementation is the reduction in number of clock cycles; the output is available after 3-clock cycle as compared to the 6-clock cycle required in standard TSPC implementation. The existence of the P block requires PMOS devices need to be sized up to attain the same gate's performance; because of its low mobility (μ_p) devices as compared to NMOS devices (μ_p).

IV. CONCLUSION

The paper presented a new modified true single-phase clock logic design style to implement the pipelined circuit for high-speed real time application. The modified implementation scheme reduces number of transistor count by 40% to 50% as compared to standard TSPC implementation.



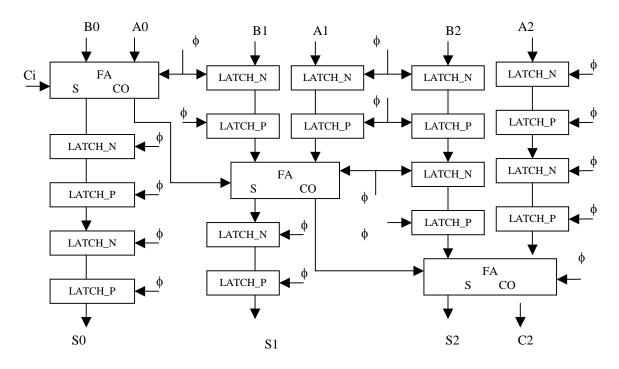


Figure 4. 3-bit pipelined adder circuit

standard TSPC				Modified TSPC			
Cell Type	No of cells	No. of transistor in each cell	Sub Total	Cell Type	No of cells	No. of transistor in each cell	Sub Total
XOR2	6	19	114	XOR2_N/XOR2_P	6	13	78
AND2	9	13	117	AND2_N/AND2_P	9	7	63
OR3	3	14	42	OR3_N/ OR3_P	3	8	24
REG	21	11	231	LATCH_N/LATCH_P	21	5	105
KLG	21	11	231	EATCH_IVEATCH_I	21	3	

TABLE IV. COMPARISON TABLE

The proposed logic style offers almost 50% reduction in the number of clock cycle therefore the pipelined logic circuit can be designed to offer reduced latency and increased throughput as compared to standard TSPC circuit implementation. The proposed modified TSPC logic design style is best suited for converting unpipelined circuits to fully pipelined circuits with effective logic merging. The circuit complexity reduces due to reduced number of transistors per logic cell. The proposed logic style implementation offers compact layout area and the reduced average power consumption for pipelined circuit implementations. Use of single clock makes clock distribution simple and compact.

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